

FREQUENCY RESPONSE ANALYSIS:

Sinusoidal signals \rightarrow steady state behaviour $\infty \rightarrow$ finite

- Characteristic eqn: $1 + GH(s) = 0$
- $\Rightarrow |GH(s)| = 1$; $\angle GH(s) = -180^\circ$ or $(2k+1)\pi$ for a just stable system \therefore Indicate Gain & Phase Margins to stability by pass. i.e. $s = j\omega$.

BODE PLOT HENDRIK W. BODE 1905

$GH(j\omega) = |GH(\omega)| e^{j\phi(\omega)}$

$\therefore \ln GH(j\omega) = \ln |GH(\omega)| + j\phi(\omega)$ [usually in dB]

log scale \therefore cover large range of freq. ; log: $\times 2$ to + and -.

(D)

$u = \log_{10} \omega \Rightarrow 10^u = \omega$

$u_2 - u_1 = \log_{10} \omega_2 - \log_{10} \omega_1 = \log_{10} \frac{\omega_2}{\omega_1}$ usually octaves $\omega_2 = 2\omega_1$, or decades $\omega_2 = 10\omega_1$

No. of decades: $\frac{\log_{10} \frac{\omega_2}{\omega_1}}{\log_{10} 10} = \log_{10} \frac{\omega_2}{\omega_1} = x \Rightarrow \frac{\omega_2}{\omega_1} = 10^x \Rightarrow \omega = 2^x$

Let $\omega = 10^u = 2^v \therefore \log_{10} \omega = u$; $\log_2 \omega = v$

$\Rightarrow v \log_{10} 2 = u \Rightarrow v = \frac{u}{\log_{10} 2} = \frac{\log_{10} \omega}{\log_{10} 2} = \log_2 \omega$

$\log_2 \frac{\omega_2}{\omega_1} = y$
No. of octaves

$\therefore 20 \text{ dB/decade} = 20 \log_{10} \frac{\omega_2}{\omega_1} = 20 \times (\log_{10} 2) \times (\log_2 \frac{\omega_2}{\omega_1}) = 20 \times 0.301 \log_2 \frac{\omega_2}{\omega_1} = 6 \text{ dB/octave}$ CHAIN RULE = 6.4

Ex $G(s) = \frac{1}{s+1} \rightarrow$ low pass RC $R=2\Omega$, $C=0.5F$ with i/p $10 \cos 4tV$.

V_0 across C?

$u(s) = \mathcal{L}[10 \cos 4t] = \frac{10s}{s^2+16} \therefore V_0(s) = \frac{10s}{(s+1)(s^2+16)} = \frac{-10/17}{s+1} + \frac{10/17s + 160/17}{s^2+16}$

(A)

Laplace

$\therefore V_0(t) = -\frac{10}{17} e^{-t} + \left(\frac{10}{17} \cos 4t + \frac{40}{17} \sin 4t \right)$

(transient) A.A.

Let $s = j\omega$, $\therefore G(j\omega) = \frac{1}{1+j\omega}$ for $\omega=4 = \frac{1-j4}{17} = \frac{1}{\sqrt{17}} \angle -76^\circ$

Sinusoidal Analysis

$\therefore V_0(t) = \frac{10}{\sqrt{17}} \cos(4t - 76^\circ) V = \frac{10}{17} \cos 4t + \frac{40}{17} \sin 4t$

Frequency Domain Analysis:

- Mag. : $20 \log |GH(s)| = 20 \log 1 = 0 \text{ dB} \rightarrow$ gives gain crossover freq. ω_c and PHASE MARGIN.
- Phase : $\angle GH(s) = (2k+1)\pi \rightarrow$ gives phase crossover freq. GAIN MARGIN.
- $= -180^\circ$

Gain Margin = $-20 \log |GH(j\omega_p)| \text{ dB} = +20 \log \left| \frac{1}{GH(j\omega_p)} \right| \text{ dB}$

Phase Margin = $\angle GH(j\omega_g) + 180^\circ \Rightarrow \angle GH(j\omega_g) - PM = -180^\circ$

(Normally lagging system \Rightarrow add phase)

Both +ve \Rightarrow STABLE system.

CONSTRUCTION OF BODE PLOT:

$$GH(s) = \frac{K_s \prod (s^2 + 2\zeta_k \omega_{nk} s + \omega_{nk}^2) \prod (sT_i + 1)}{s^N \prod (s^2 + 2\zeta_k \omega_{nk} s + \omega_{nk}^2) \prod (sT_e + 1)}$$

$$GH(j\omega) = \frac{K \prod (j\omega T_i + 1) \prod (1 + j2\zeta_k \omega_{nk} - \omega_{nk}^2)}{(j\omega)^N \prod (j\omega T_e + 1) \prod (1 + j2\zeta_k \omega_{nk} - \omega_{nk}^2)}$$

where $K = \frac{K_s \prod \omega_{nk}^2}{\prod \omega_{nk}^2}$ $\omega_{nk} = \frac{\omega}{\omega_{nk}}$ $\omega_{nk} = \frac{\omega}{\omega_{nk}}$

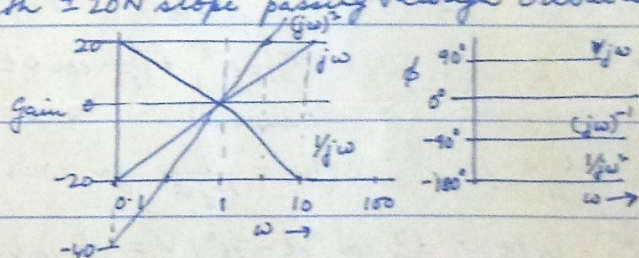
Mag: $|GH(j\omega)|_{dB} = 20 \log |K| + \sum 20 \log |1 + j\omega T_i|$
 $+ \sum 20 \log |1 + j2\zeta_k \omega_{nk} - \omega_{nk}^2|$
 $- 20N \log |\omega| - \sum 20 \log |1 + j\omega T_e|$
 $- \sum 20 \log |1 + j2\zeta_k \omega_{nk} - \omega_{nk}^2|$

Phase: $\angle GH(j\omega) = 0 + \sum \tan^{-1}(\omega T_i) + \sum \tan^{-1} \frac{2\zeta_k \omega_{nk}}{1 - \omega_{nk}^2}$
 $- N \cdot \frac{\pi}{2} - \sum \tan^{-1}(\omega T_e) - \sum \tan^{-1} \frac{2\zeta_k \omega_{nk}}{1 - \omega_{nk}^2}$

① Constant K $20 \log |K| = \text{constant}$; phase = 0.

② Zeros/poles at origin $(j\omega)^{\pm N}$ \rightarrow ② 20dB/dec = 6dB/octave
 $\pm 20N \log(j\omega) = \pm 20N \log|\omega| \text{ dB}$

- straight line with $\pm 20N$ slope passing through 0dB at $\omega=1$.
 angle of $\pm N (\pi/2)$



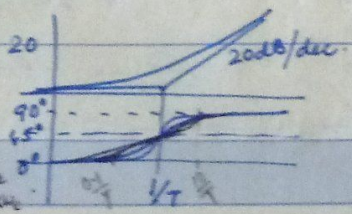
③ Simple zero/pole $(1+j\omega T)^{\pm 1}$ \rightarrow Extra sheet

Let $GH(j\omega) = 1 + j\omega T$ \therefore Mag: $20 \log \sqrt{1 + \omega^2 T^2}$
 For $\omega T \ll 1$, Mag = 0 $\omega T \gg 1$, Mag = $20 \log \omega + 20 \log T$

st. line with slope 20dB/dec
 with origin at $1/T = \omega$.

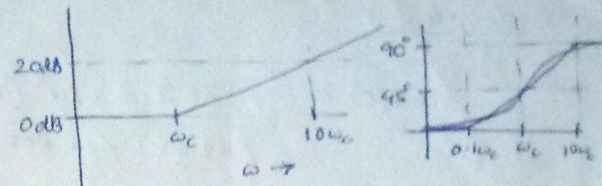
Phase: $\tan^{-1} \omega T$ \therefore 45° at $\omega T = 1$ or $\omega = 1/T$

Let $\omega = K \omega_c$, $\omega > \omega_c$
 $\omega = \frac{1}{K} \omega_c$, $\omega < \omega_c$
 Both cases,
 Error = $20 \log \sqrt{\frac{1+K^2}{K^2}}$
 $\omega T = \frac{\omega}{\omega_c} = \frac{K \omega_c}{\omega_c} = K$



Error: $20 \log \sqrt{1 + \left(\frac{0.5 \omega_c}{\omega_c}\right)^2} = (20 \log 1.2) = 1.8 \text{ dB}$
 $\approx 1 \text{ dB}$ at $0.5 \omega_c, 2 \omega_c$
 $\approx 3 \text{ dB}$ at ω_c
 $\approx 0.04 \text{ dB}$ at $0.1 \omega_c, 10 \omega_c$
 $L: \pm 5.7^\circ$ at $0.1/\omega_c$; $\pm 2.9^\circ$ at $0.5/\omega_c$ for zero at $(1+j\omega T)$

Simple zero and pole $(1+j\omega T)^{\pm 1}$
 det $GH(j\omega) = 1+j\omega T$ det $T = 1/\omega_c$
 $= 1+j\left(\frac{\omega}{\omega_c}\right)$



Mag: $20 \log \sqrt{1+\omega^2 T^2} = 20 \log \sqrt{1+\left(\frac{\omega}{\omega_c}\right)^2}$ dB

$\omega T = \frac{\omega}{\omega_c} \ll 1$ $20 \log 1 = 0$ dB
 $\omega T = 1$ $20 \log 2 = 3.01$ dB
 $\omega T \gg 1$ $20 \log \left(\frac{\omega}{\omega_c}\right) = 20 \log \omega + 20 \log T$

Phase $\tan^{-1} \omega T = \tan^{-1} \frac{\omega}{\omega_c}$
 $\tan^{-1} 0 = 0^\circ$
 $\tan^{-1} 1 = 45^\circ$
 $\tan^{-1} \frac{\omega}{\omega_c} \rightarrow \tan^{-1} \infty = 90^\circ$

For $\frac{\omega}{\omega_c} = K$, $20 \log K$; $K=10 \Rightarrow 20$ dB.
 $\therefore 20$ dB/decade slope, origin at $\omega = \omega_c = \frac{1}{T}$

Approx. 0° upto $0.1/\omega_c$
 90° beyond $10\omega_c$
 st. line from $0.1/\omega_c$ to $10\omega_c$
 with 45° at ω_c .
 $y = 45 \left[\log \frac{\omega}{\omega_c} + 1 \right]$ at $\omega = \omega_c$
 $\log 1 = 0$
 \therefore approx. origin

Error: $20 \log \sqrt{1+\left(\frac{\omega}{\omega_c}\right)^2}$ for $\omega \ll \omega_c$
 $20 \log \sqrt{1+\left(\frac{\omega}{\omega_c}\right)^2} - 20 \log \frac{\omega}{\omega_c}$ for $\omega > \omega_c$

Phase error $= 45 \left[\log \frac{\omega}{\omega_c} + 1 \right]$
 $\tan^{-1} \omega T - y$

Let $\frac{\omega}{\omega_c} = K$ for $\omega > \omega_c$ \therefore Error $= 20 \log \sqrt{\frac{1+K^2}{K^2}}$
 - do -
 For $\frac{\omega}{\omega_c} = \frac{1}{K}$

$\pm 5.71^\circ$ | Phase error 0 at approx $K = 0.1588 = 1/\sqrt{2}$
 $\mp 4.89^\circ$
 0

$\therefore K = 0.1 \text{ or } 10$ $+ 0.043$ dB
 $K = 0.5 \text{ or } 2$ $+ 0.97 \approx 1$ dB
 $K = 1$ $+ 3.01$ dB

Quadratic zero and pole: Mag $\propto \frac{1}{g(\omega)}$; At $\omega = \omega_n$, $G(j\omega_n) = \frac{1}{2\xi}$

$g(\omega) = \left[\frac{\omega^2 - \omega_n^2(1-2\xi^2)}{\omega_n^2} \right]^2 + 4\xi^2(1-\xi^2)$ $\omega \ll \omega_n \rightarrow 1$

For $\xi = 1$ $g(\omega) = \left[\frac{\omega^2 + \omega_n^2}{\omega_n^2} \right]^2 > 1 \rightarrow 1$

$\xi = \frac{1}{\sqrt{2}} = 0.707$ $\left[\frac{\omega^2}{\omega_n^2} \right]^2 + \frac{4}{2} \cdot \frac{1}{2} = 1 + \left(\frac{\omega}{\omega_n}\right)^2 \rightarrow 1$

$\xi = \frac{1}{2} = 0.5$ $\left[\frac{2\omega^2 - \omega_n^2}{2\omega_n^2} \right]^2 + \frac{4}{4} \cdot \frac{3}{4} \rightarrow 1$

$\xi \ll 1$
 $\xi = 0.05$ $\left[\frac{\omega^2 - \omega_n^2(0.99975)}{\omega_n^2} \right]^2 + 0.0005 \rightarrow 1$

$\omega_n = \omega_n \sqrt{1-2\xi^2}$ for resonant ω

$\omega = \omega_n$	$G(j\omega_n)$	$\omega = \omega_n$	Value of ω_n
$4 \rightarrow \infty$	$\frac{1}{2}$	NA	Not valid
$2 - j\sqrt{2} \rightarrow \infty$	1	0	0
$1 - j \rightarrow \infty$	0.75	0.75	$\frac{\omega_n}{\sqrt{2}}$
$0 \rightarrow \infty$	0	0	ω_n
$0.0005 - 10j \rightarrow \infty$	0.0005	0.0005	$0.999875 \omega_n$

$\times \left[\left(\frac{\omega^2}{\omega_n^2}\right)^2 - 2(1-2\xi^2) \frac{\omega^2}{\omega_n^2} + (1-2\xi^2)^2 - (1-2\xi^2)^2 \right]^2$
 $= \left[\frac{\omega^2 - \omega_n^2(1-2\xi^2)}{\omega_n^2} \right]^2 - 4\xi^2(1-\xi^2)$

At ω_r , $\frac{\omega_r^2}{\omega_n^2} = \sqrt{1-2\xi^2} \Rightarrow GH(j\omega_r) = \frac{1}{1 - (1-2\xi^2) + j2\xi\sqrt{1-2\xi^2}}$
 $= \frac{1}{2\xi(\xi + j\sqrt{1-2\xi^2})} \therefore |GH(j\omega_r)| = \frac{1}{2\xi\sqrt{1-\xi^2}}$

At $\omega_r = \omega_n = \omega_n \sqrt{1-2\xi^2}$
 $\omega = \frac{\omega}{\omega_n} = \frac{\omega_n \sqrt{1-2\xi^2}}{\omega_n \sqrt{1-2\xi^2}} = 1$
 $\therefore GH(j\omega_r) = \frac{1}{1 - 2\xi^2 + j2\xi\sqrt{1-2\xi^2}}$
 $= \frac{1-2\xi^2}{2\xi(-\xi + j\sqrt{1-2\xi^2})}$

Gain Margin = $20 \log \frac{1}{|G H(j\omega_c)|}$ dB at phase crossover freq.
 (by what factor can gain be inc.)

Phase Margin = $\angle G H(j\omega_g) + 180^\circ$

∴ If stable, then G.M. > 1 , PM +ve (amount of phase lag allowable that pushes sys. to instability).
 (multiplying factor) > 0 dB

→ ω_{ref} for pf
 * Loading - // paths - i_x increases
 ∴ loading in terms of i
 ∴ Inductive load lagging pf. Lags 'LOAD' or 'i' w.r.t v .
 usually inductive or lagging loads - coils, \dots

1. Identify critical frequencies, magnitude & phase at these frequencies
2. Plot bode components while stating characteristics
3. Total plot — additive : semi log plot for magnitude & phase
4. Identify $\omega_g, \omega_p, G.M., P.M.$ from plot. Verify with tabular values
 Comment — % error, why & expectations.
5. Infer \underline{Q} stability.

f	M_H	$\angle H$

$$\frac{K(s+50)}{s(s+20)(s^2+800s+64 \times 10^4)}$$

Bode plot :

$$GH(s) = \frac{13653.3 \times 10^8 (s+150)}{s(s+40)(s+800)(s^2+2000s+4 \times 10^6)}$$

$$\therefore GH(j\omega) = \frac{1600 \left(1 + \frac{j\omega}{150}\right)}{j\omega \left(1 + \frac{j\omega}{40}\right) \left(1 + \frac{j\omega}{800}\right) \left[1 - \left(\frac{\omega}{2000}\right)^2 + \frac{j\omega}{2000}\right]}$$

Four factors:

- 1) Constant $K = 1600 \triangleq 64 \text{ dB} = (20 \log 1600)$: shift of origin, 0° phase
- 2) Pole at origin : $(j\omega)^{-1}$: Mag : -20 dB/dec through $\omega=1$, phase = -90°
- 3) Simple poles and zeros :

a) $\left(1 + j\frac{\omega}{40}\right)^{-1}$		Mag		Phase
b) $\left(1 + j\frac{\omega}{150}\right)$		0dB till 40, then -20 dB/dec		0 till 4, -45° at 40, -90° from 400
c) $\left(1 + j\frac{\omega}{800}\right)^{-1}$		150, $+20 \text{ dB/dec}$		till 15, 45° at 150, 90° from 1500
		800, -20 dB/dec		till 80, -45° at 800, -90° from 8000
- 4) Quadratic pole : $1 - \left(\frac{\omega}{2000}\right)^2 + \frac{j\omega}{2000}$: Asymptote : 0 till 2000, -40 dB/dec then
 $\therefore \xi = 0.5$
 $G(j2000) = \frac{-1}{1} = -1$ Mag 1
 $\omega_n = \omega_c \sqrt{1 - 2\xi^2} = \frac{2000}{\sqrt{2}} = 1414$
 $|G(j\omega_n)| = \frac{1}{\sqrt{\frac{10}{16}}} = \frac{4}{\sqrt{10}} = 1.265 = 2.04 \text{ dB}$

Phase : 0° till 200, -90° at 2000, -180° from 20000

Critical Freq.	Mag. (Th.)	Ph. (Th.)
1		
4		
15		
40		
80		
150		
200		
400		
800		
1500		
2000		
8000		
20000		
	GM +12 dB	-180° ω_p (1150 Hz)
	1 ω_g (420 Hz)	PM (44°)

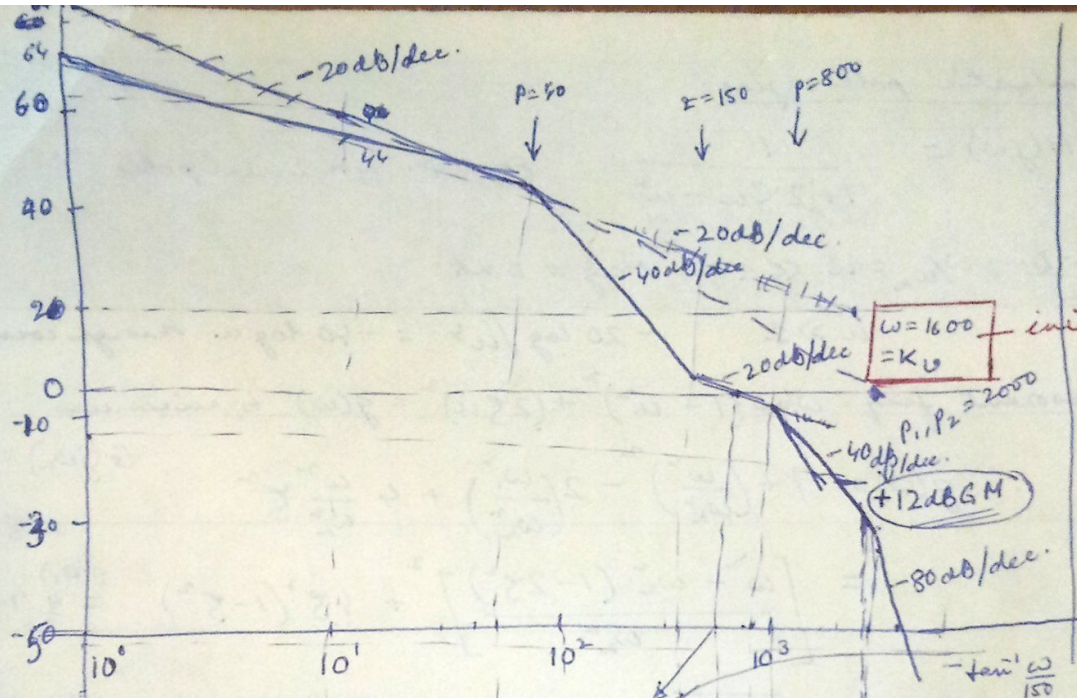
Theo.

Comparison of Theo. and Practical ω_p , GM, ω_g , PM.

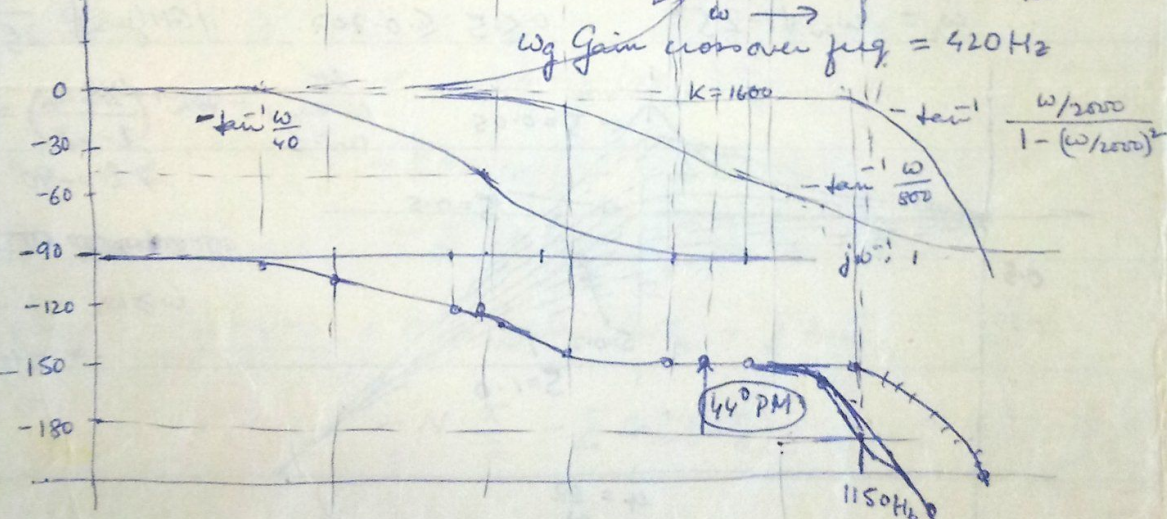
Gain Margin : $20 \log \frac{1}{|GH(j\omega_p)|}$ dB at ω_p : If stable, then +ve GM ($\because \frac{1}{|GH(j\omega_p)|} > 1$)

Phase Margin : $|GH(j\omega_g)| - (-180^\circ)$: If stable, then +ve PM (amount of log allowable)

Infer & stability.



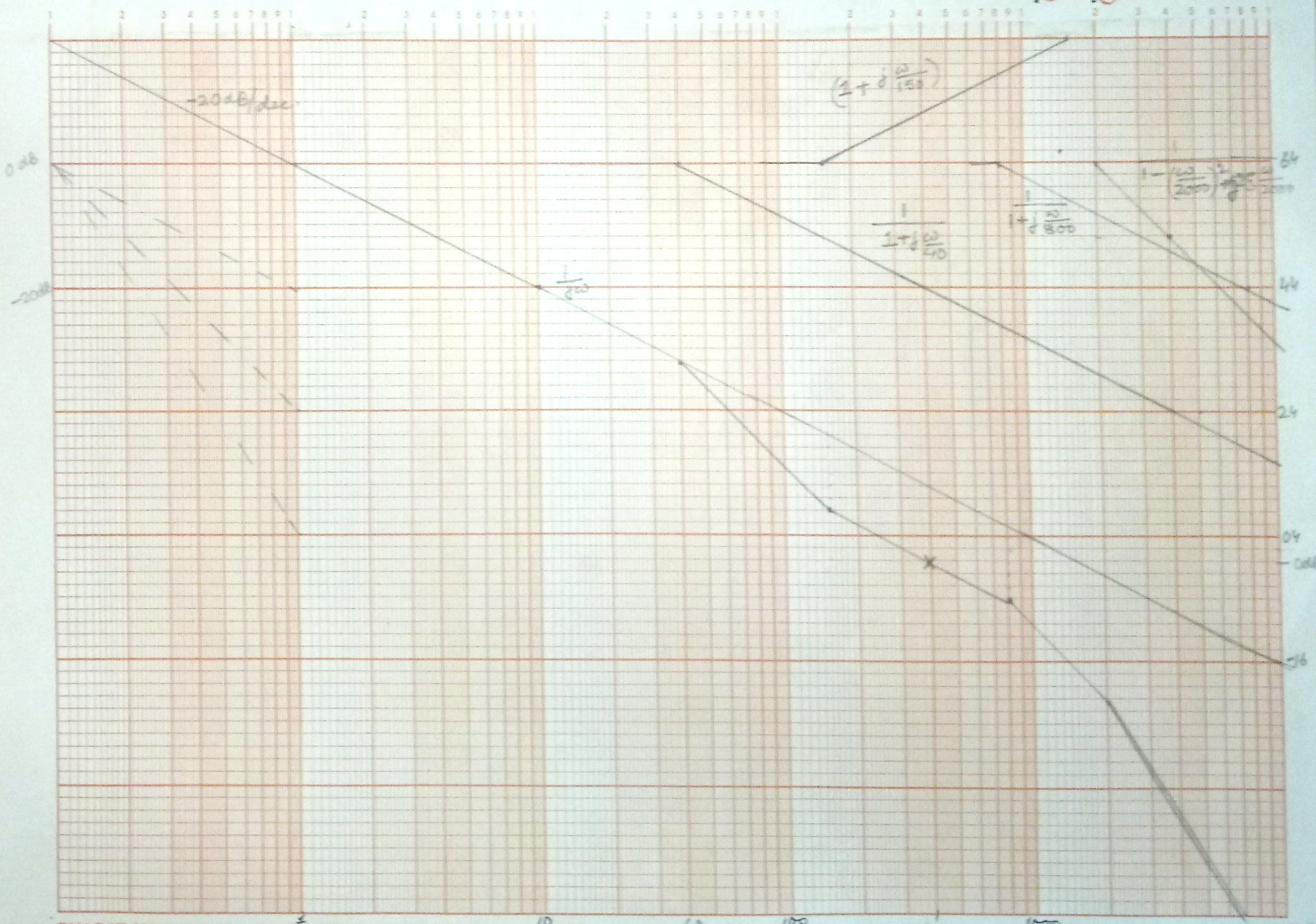
$0 = -20x + 64$
 $\Rightarrow x = 3.2 \Rightarrow \omega_c = 10 \cdot \omega_c^{3.2}$
 initial slope of $= 1585 \text{ rad/s}$
 20 dB/dec.
 intersection with 0 dB.
 OR $\omega = 1, K = K_0 = 1600$
 $= \lim_{s \rightarrow 0} sG(s)$



ω_p phase crossover freq.

14
21/13

Shalby



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Magnitude

smooth curve through the calculated points. See Figure 8.11.

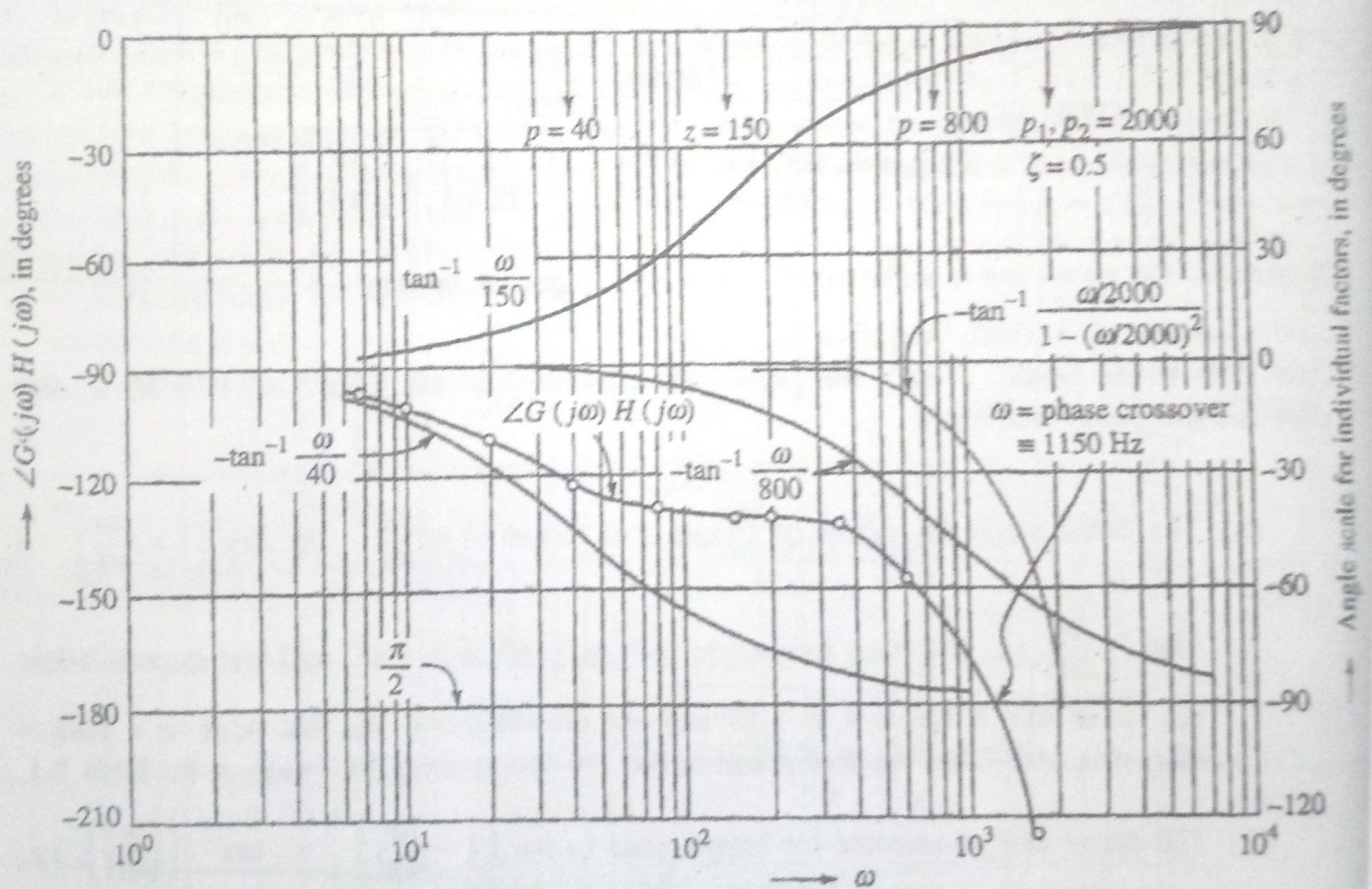


FIGURE 8.11 Example 8.4: phase angle curve.

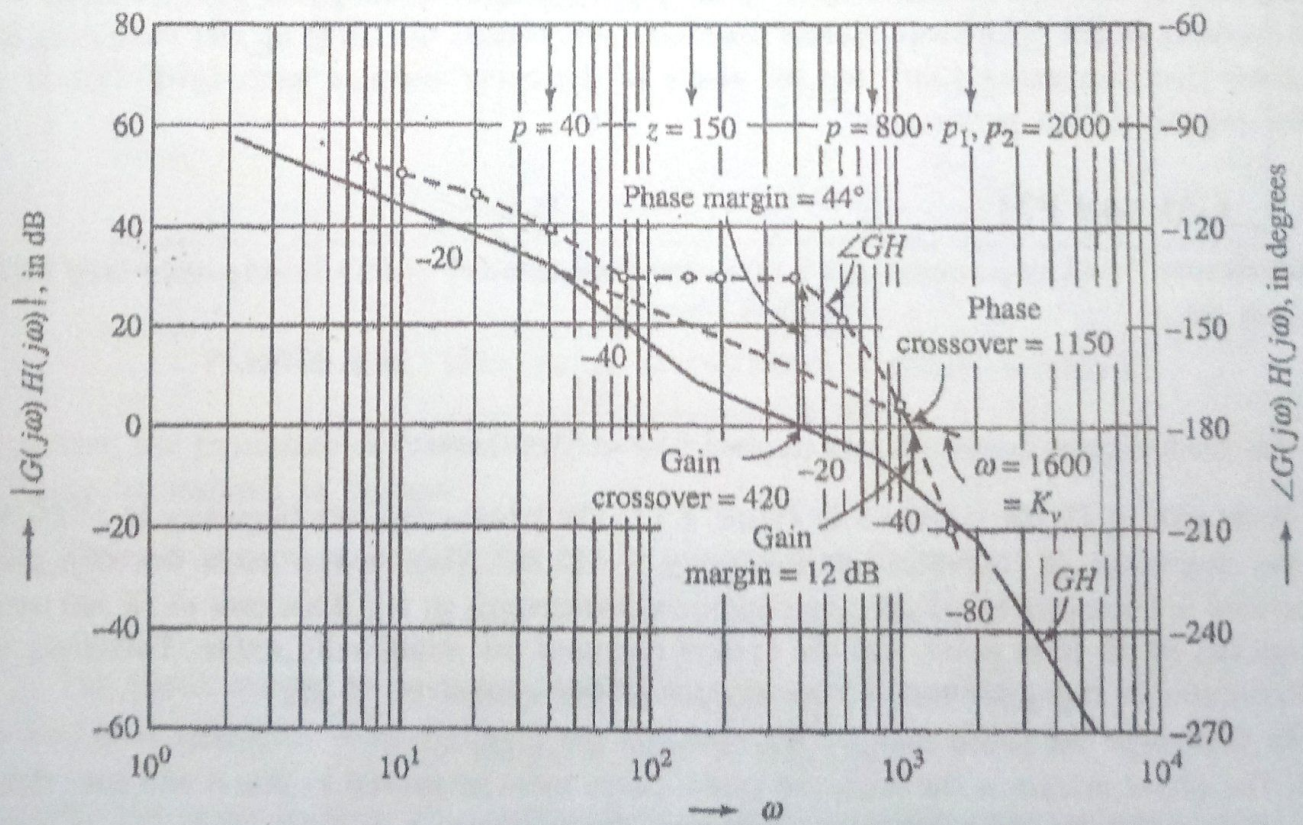


FIGURE 8.12 Example 8.5: magnitude and phase curves.