

Polar plot (NYQUIST PLOT) $\omega: 0 \text{ to } \infty$ (18)

plot of $|G(j\omega)|$ vs $\angle G(j\omega)$ in polar co-ord.

\Rightarrow $\text{Re } G(j\omega)$ vs $\text{Im } G(j\omega)$ in Cartesian co-ord.

$|G(j\omega)|$ and $\text{Re } G(j\omega)$ are even fns.

$\angle G(j\omega)$ and $\text{Im } G(j\omega)$ are odd fns.

+ve phase angle is CCW.

Disadv.: a) addn of poles & zeros to existing sys. req. recalculations
 similar to root locus b) effects of addn. of individual poles/zeros not indicated

For cascaded T.F., use Bode plot & shift to polar plot.

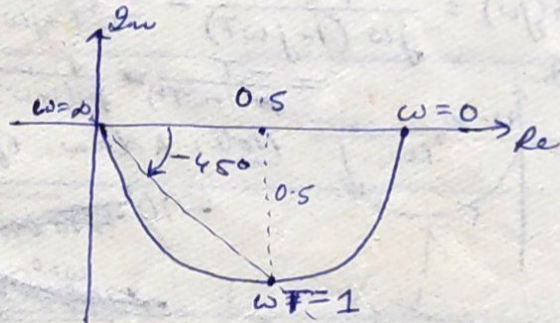
Adv. freq. response over entire freq. range in one plot.

Ex Low pass RC (LAG) ckt.

$$G(s) = \frac{V_2(s)}{V_1(s)} = \frac{1}{1+sT}; \quad T=RC$$

$$\therefore G(j\omega) = \frac{1}{1+j\omega T} = \frac{1}{\sqrt{1+\omega^2 T^2}} \angle -\tan^{-1} \omega T$$

ω	$ G(j\omega) $	$\angle G(j\omega)$
0	1	0
$1/T$	0.707	-45°
∞	0	-90°

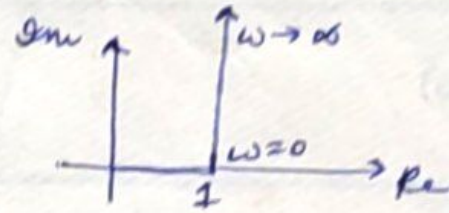


\therefore Polar plot is a semicircle of rad. 0.5 and center at 0.5.

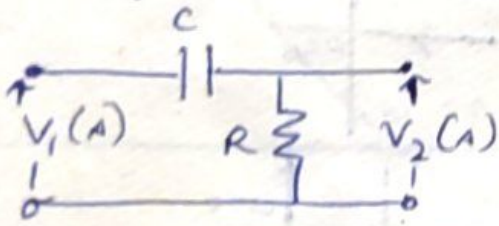
$$(x - 1/2)^2 + y^2 = (1/2)^2 \quad \text{where } x = \text{Re } G(j\omega) = \frac{1}{1+\omega^2 T^2}, \quad y = \text{Im } G(j\omega) = \frac{-\omega T}{1+\omega^2 T^2}$$

0/p lags behind i/p: indication is plot in $(-j\omega)$ plane.

Ex For $G(j\omega) = 1 + j\omega T \rightarrow$ LEADS op.
 ex. series RL ckt.
 (NON CRUSAL)

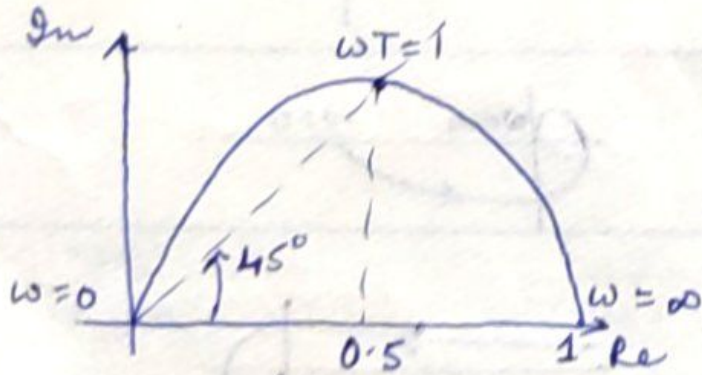


Ex High pass RC (LEAD) Network



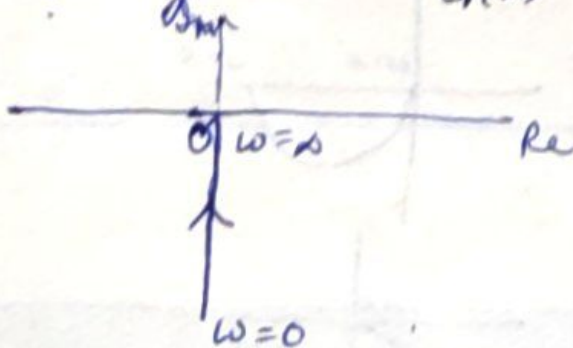
$$G(s) = \frac{V_2(s)}{V_1(s)} = \frac{s}{s + 1/T} \quad T = RC.$$

$$\Rightarrow G(j\omega) = \frac{j\omega}{j\omega + 1/T} = \left| \frac{\omega T}{\sqrt{1 + \omega^2 T^2}} \right| \angle 90^\circ - \tan^{-1} \omega T$$



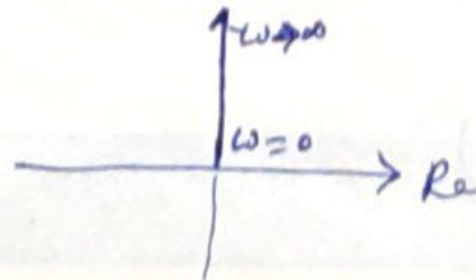
opp voltage leads up at all times.

Ex Integrator $G(s) = 1/s$



$$\therefore G(j\omega) = \frac{1}{j\omega} = \frac{-j}{\omega} = \frac{1}{\omega} \angle -90^\circ$$

Ex Derivative factor $G(s) = s$



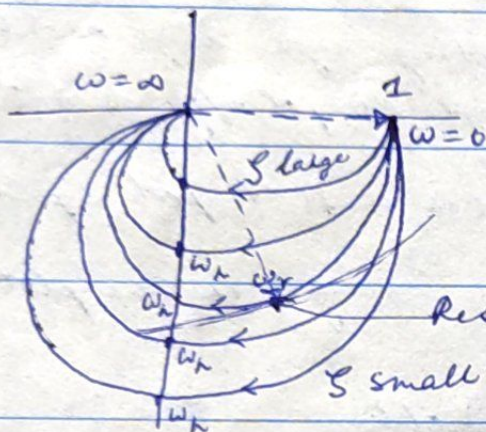
Ex. RLC series network with opp across capacitor

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$\omega_n = \frac{1}{\sqrt{LC}}, \quad \zeta = \frac{R}{2}\sqrt{\frac{L}{C}}$$

$$\therefore G(j\omega) = \frac{1}{1 + j2\zeta(\omega/\omega_n) - (\omega/\omega_n)^2}$$

ω	$G(j\omega)$
0	$1 \angle 0^\circ$
∞	$0 \angle 180^\circ \rightarrow -\tan^{-1}\left(\frac{2\zeta}{(-\omega/\omega_n)}\right) \rightarrow -\tan^{-1}(0)$



At $\omega = \omega_n$, $G(j\omega_n) = \frac{1}{j2\zeta} = -j \frac{1}{2\zeta}$

\therefore on neg. real axis

ω_r where distance from origin is max. $\omega_r = \omega_n \sqrt{1 - 2\zeta^2}$

$$|G(j\omega_r)| = \frac{1}{2\zeta \sqrt{1 - \zeta^2}}$$

Resonant mag. = $\frac{1}{|\text{vector}(\omega_r)|}$

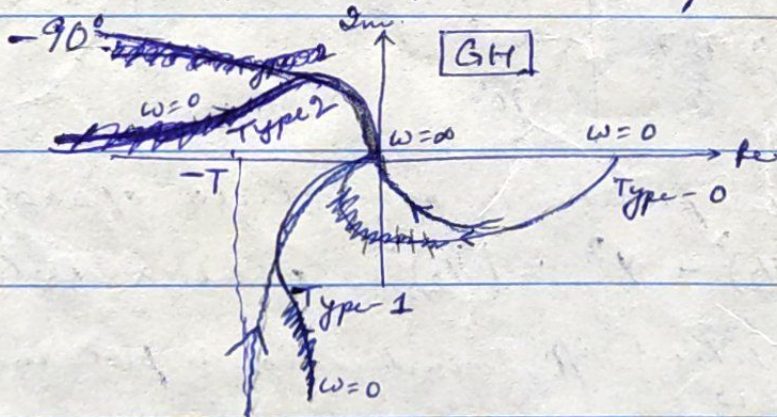
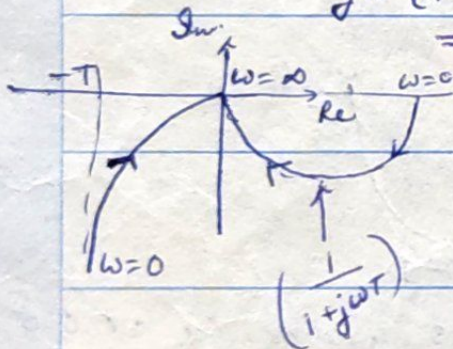
ex $G(s) = \frac{1}{s(sT+1)}$

$$G(j\omega) = \frac{1}{j\omega(1+j\omega T)} = \frac{1}{-j\omega T + j\omega}$$

$$= \frac{-1}{(1+\omega^2 T^2)} = \frac{j}{j\omega(1+\omega^2 T^2)}$$

ω	$G(j\omega)$
0	$-T - j\infty = \infty \angle -90^\circ$
∞	$0 \angle -180^\circ$

Note: Addn. of simple pole rotates polar plot by



$$G(j\omega) = \frac{1}{(j\omega)^2(1+j\omega T)}$$

$$= \frac{-j}{\omega} \left[\frac{-T}{1+\omega^2 T^2} - \frac{j}{\omega(1+\omega^2 T^2)} \right]$$

$$= \frac{-1}{\omega^2(1+\omega^2 T^2)} + j \frac{T}{\omega(1+\omega^2 T^2)}$$

$$= -\infty + j\infty \text{ at } \omega=0.$$

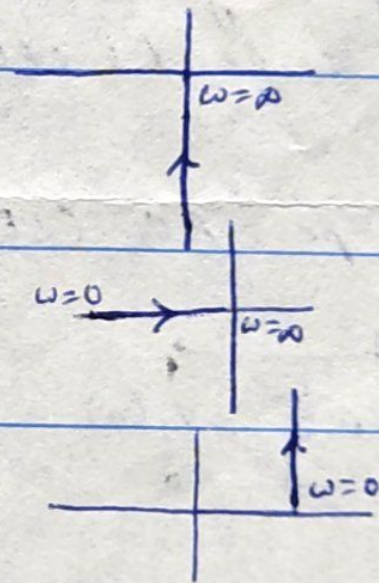
$$G(j\omega)$$

Pole-zero

Polar plot

Rem.

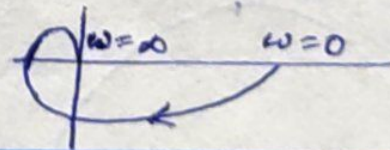
$$\frac{1}{j\omega}$$



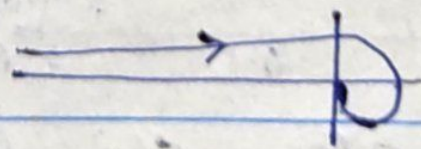
$$\frac{1}{(j\omega)^2}$$

$$1 + j\omega T$$

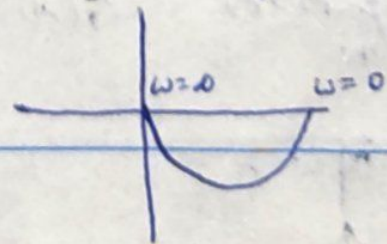
$$\frac{1}{(1 + j\omega T_1)(1 + j\omega T_2)(1 + j\omega T_3)}$$



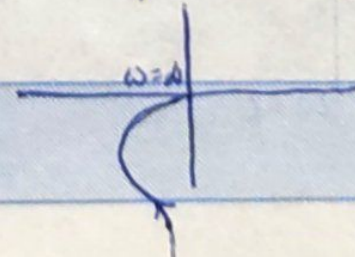
$$\frac{1}{(j\omega)^2 (1 + j\omega T_1)(1 + j\omega T_2)(1 + j\omega T_3)}$$



$$\frac{1}{1 + j\omega T}$$

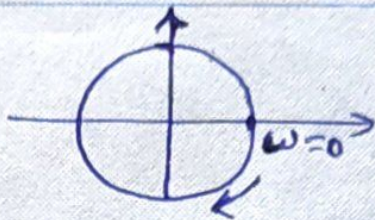


$$\frac{1}{j\omega (1 + j\omega T)}$$



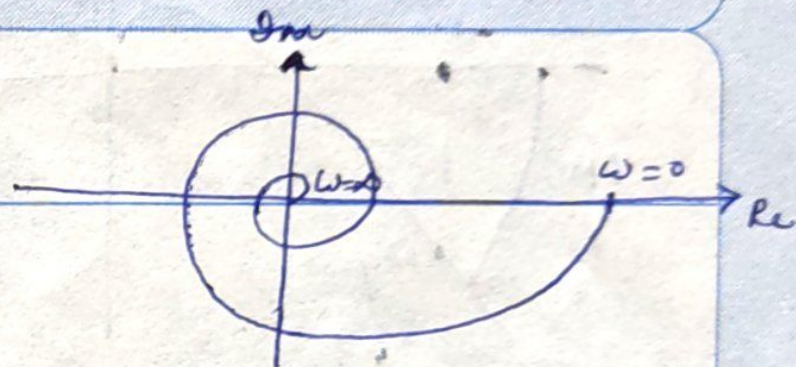
Ex Transportation lag $G(s) = e^{-sT}$

$$G(j\omega) = 1 \angle -\omega T$$



Ex $G(s) = \frac{\exp(-sL)}{1+sT}$

$$\therefore G(j\omega) = \frac{1}{\sqrt{1+\omega^2 T^2}} \angle -\omega L - \tan^{-1} \omega T$$



Ex $G(s) = \frac{10}{s(s+1)(s+2)}$

$$G(j\omega) = \frac{10(-j\omega)(1-j\omega)(2-j\omega)}{\omega^2(1+\omega^2)(2+\omega^2)}$$

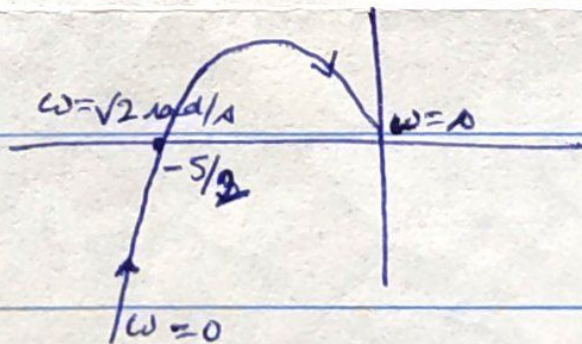
ω	$G(j\omega)$
0	$\infty \angle -90^\circ$
∞	$0 \angle -270^\circ$
$\sqrt{2}$	$\frac{5}{2} \angle -180^\circ$

$$\text{Re } G(j\omega) = 0 \text{ at } \omega = \infty$$

$$\text{Im } G(j\omega) = 0 \text{ at } \omega^2 = 2$$

$$= \frac{10[-3\omega^2 - j\omega(2-\omega^2)]}{\omega^2(1+\omega^2)(2+\omega^2)}$$

$$\text{Re } G(j\omega) \Big|_{\omega=\sqrt{2}} = \frac{10}{2(3)(4)} = -\frac{5}{2}$$

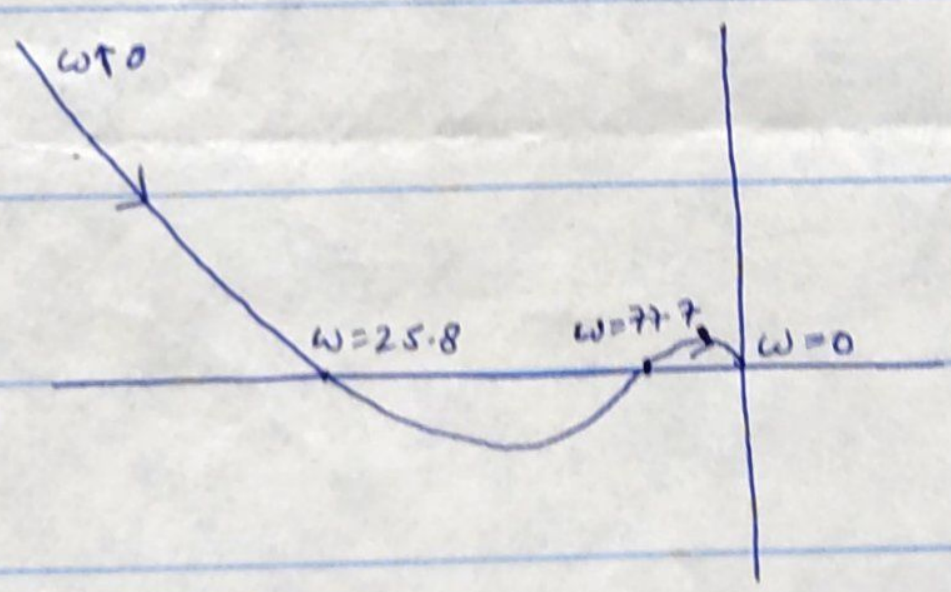


$\omega = 0$

Ex $G_H(s) = \frac{100K (s+5)(s+40)}{s^3 (s+100)(s+200)}$

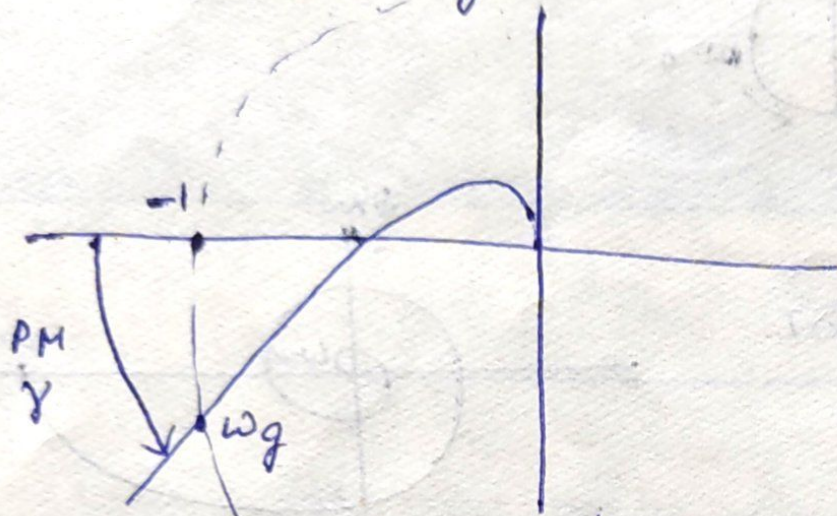
ω	$ G_H(j\omega) $	$\angle G_H(j\omega)$
0	∞	-270°
$0 < \omega < 25.8$	finite	$-180^\circ < \theta < -270^\circ$
25.8	$(-3.497 \times 10^4)K$	-180°
	finite	$-180^\circ < \theta < -90^\circ$
77.7	$(-5.3 \times 10^{-5})K$	-180°
∞	0	-270°

$\text{Im } G_H(j\omega) = 0$
 at $\omega = 25.8 \text{ rad/s}$
 & 77.7 rad/s

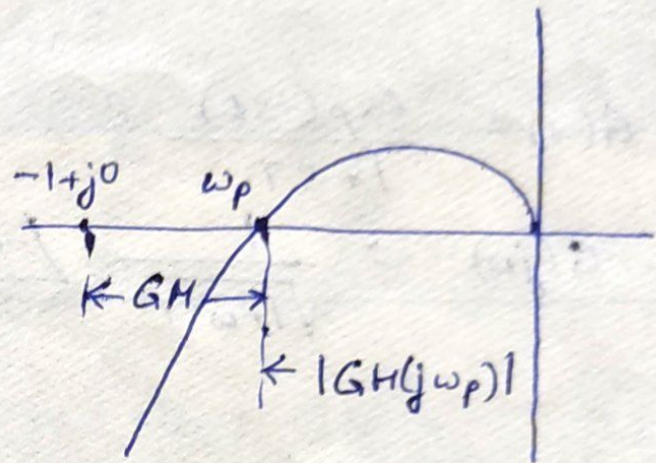


Relative stability : GM & PM.

Phase Margin



Gain Margin



$\therefore GM = 20 \log_{10} \frac{1}{|GH(j\omega_p)|} \text{ dB.}$

For MP systems, both GM & PM +ve \Rightarrow ξ stable SUFFICIENT
CONDN.

NYQUIST STABILITY:

(19)

Basic philosophy:

$$\frac{C(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$

$$F(s) = 1 + G(s)H(s) = 0 \quad ; \quad GH(s) = \frac{N(s)}{D(s)}$$
$$= 1 + \frac{N(s)}{D(s)} = \frac{N'(s)}{D(s)}$$

∴ ① \mathcal{Z} poles of system = Zeros of $F(s)$ = ROOTS of charac. eqn.

② \mathcal{Z} T.F. $GH(s)$ poles = poles of $F(s)$

③ For asymptotic stability of \mathcal{Z} sys.,

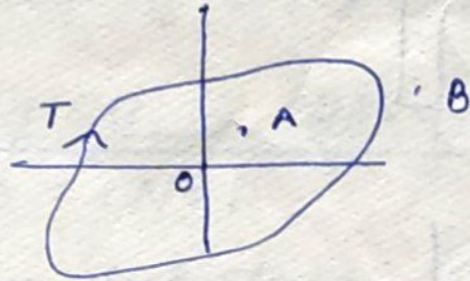
NO restriction on poles/zeros of \mathcal{Z} T.F. $GH(s)$

BUT Poles of \mathcal{Z} T.F. = Zeros of $F(s)$ in LHS.

Finding zeros of $F(s)$ is difficult but using $GH(s)$ roots, infer stability.

ANALYTIC FN $F(s)$ in s -plane if f_n & ~~all~~ all its derivatives exist \Rightarrow no singular pts. \Rightarrow no POLES.

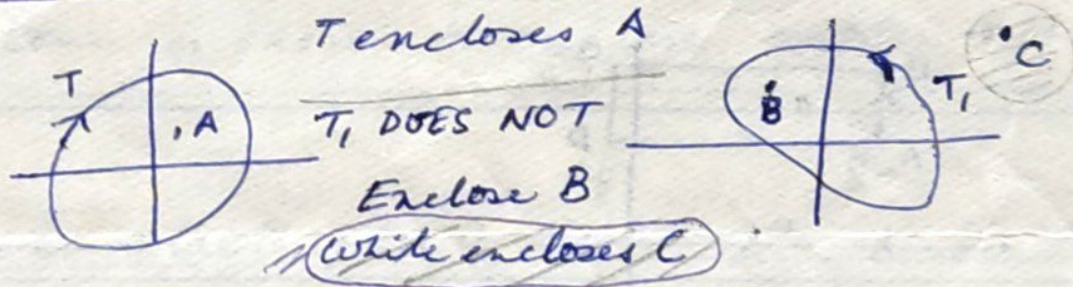
ENCIRCLED : A pt. is encircled by a Closed path if it is found inside the path. Usually CW (convention for +ve traversal)



O, A encircled
B not.

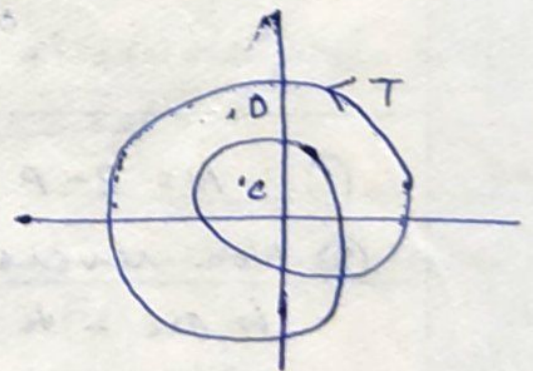
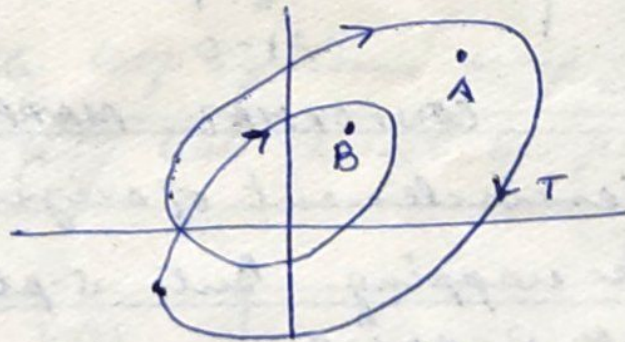
ENCLOSED

A point or region is said to be enclosed by a closed path if it is on the RT. of a path when the path is traversed in CW direction.



No. of encirclements

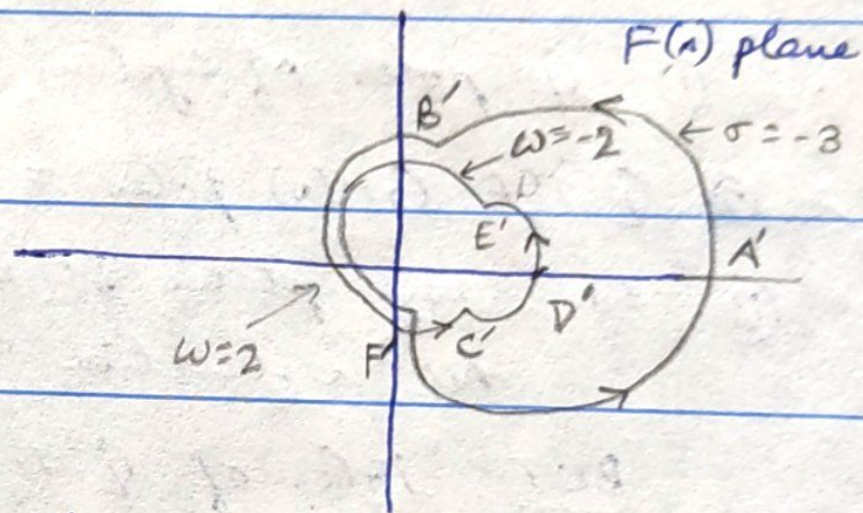
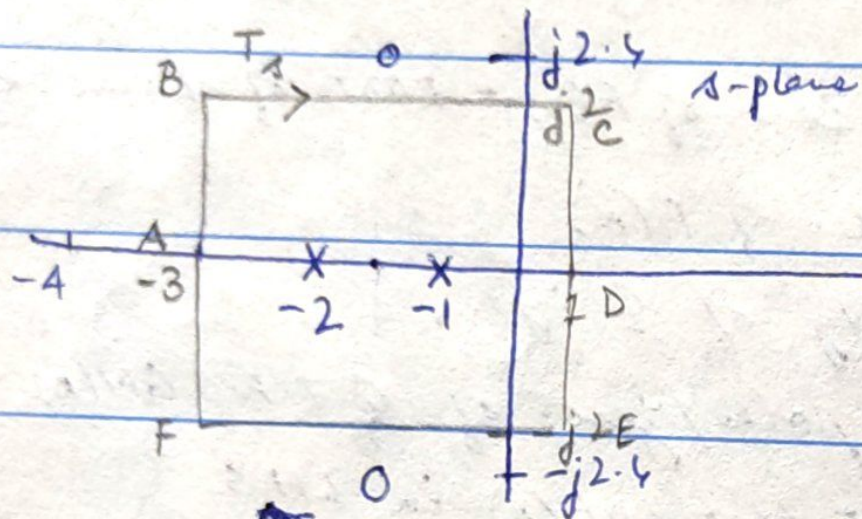
- For B $N = +2$
- A $N = +1$
- C $N = -2$
- D $N = -1$



Conformal mapping:

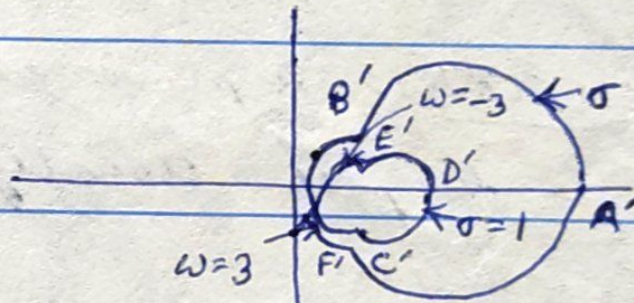
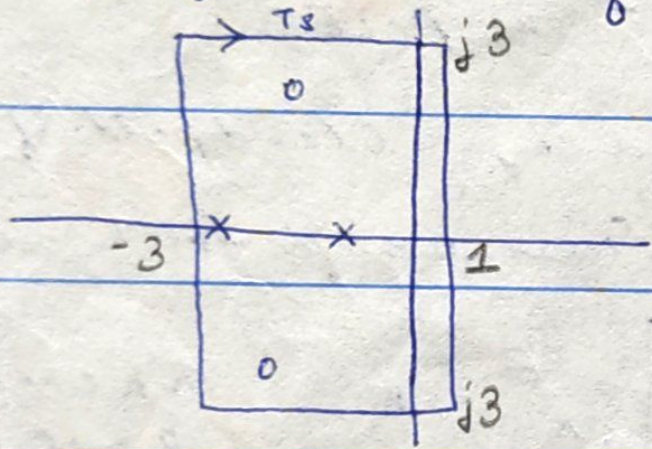
Say $GH(s) = \frac{6}{(s+1)(s+2)}$

$\therefore F(s) = 1 + GH(s) = \frac{(s+1.5+j2.4)(s+1.5-j2.4)}{(s+1)(s+2)} = 0$

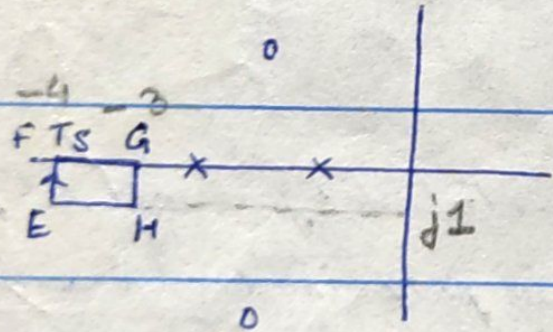


$s = \sigma + j\omega$	A = -3 + j0	B = -3 + j2	C = 1 + j2	D = 1 + j0	E = 1 - j2	F = -3 - j2
$F(s) = u + jv$	4 + j0	0.7 + j0.9	1.11 - j0.577	2 + j0	1.11 + j0.577	0.7 - j0.9

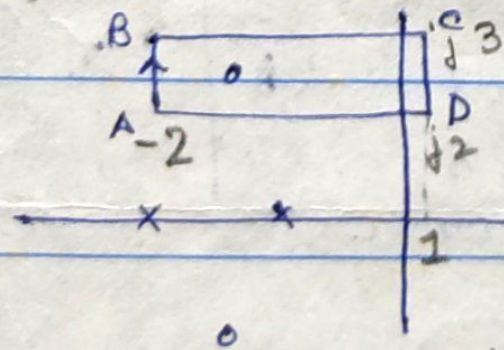
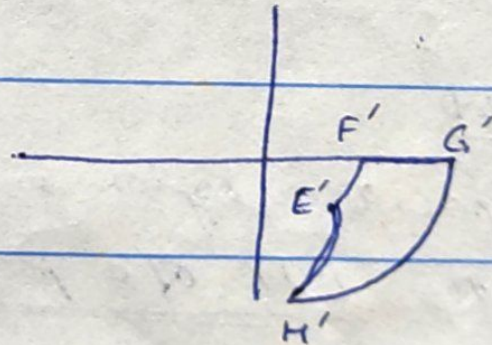
∴ No. of encirclements of origin $N = -2 = \boxed{Z - P}$ as in $T_s \rightarrow$ hence can be used as CHECK for ANALYTICAL FN. in specified region



$$N = 0 = Z - P = 2 - 2$$



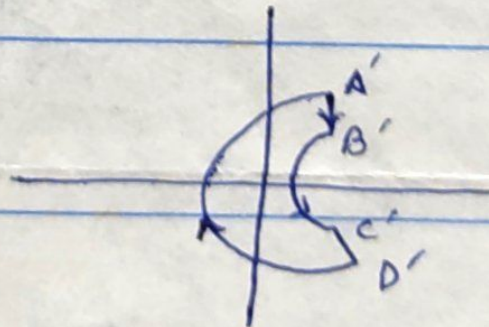
$$N = 0$$



$$N = +1$$

$$= Z - P$$

$$= 1 - 0$$



CONFORMAL MAPPING.

II

III

IV

s

F(s)

s

F(s)

s

F(s)

$$A = -3 + j0$$

$$4 + j0$$

$$A = -4 + j0$$

$$2 + j0$$

$$A = -2 + j2$$

$$-0.2 + j0.6$$

$$B = -3 + j3$$

$$0.68 + j0.41$$

$$B = -3 + j0$$

$$4 + j0$$

$$B = -2 + j3$$

$$0.4 + j0.2$$

$$C = 1 + j3$$

$$0.92 - j0.38$$

$$C = -3 + j1$$

$$1.6 - j1.8$$

$$C = 1 + j3$$

$$0.92 - j0.38$$

$$D = 1 + j0$$

$$2 + j0$$

$$D = -4 - j1$$

$$1.6 - j0.6$$

$$D = 1 + j2$$

$$1.11 - j0.58$$

$$E = 1 - j3$$

$$0.92 + j0.38$$

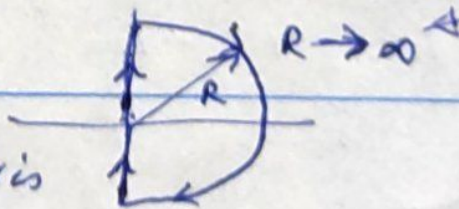
$$F = -3 - j3$$

$$0.68 - j0.41$$

CONFORMAL MAPPING.

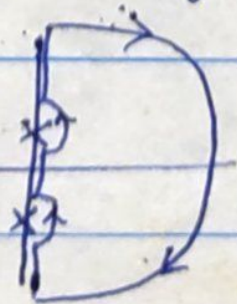
- ① $\therefore N = Z - P$ (encirclement of origin)
- ② Non reversible mapping But: s -plane to $F(s)$ plane is one-to-one mapping

Usually T_s as $j\omega$ axis to RHS -plane



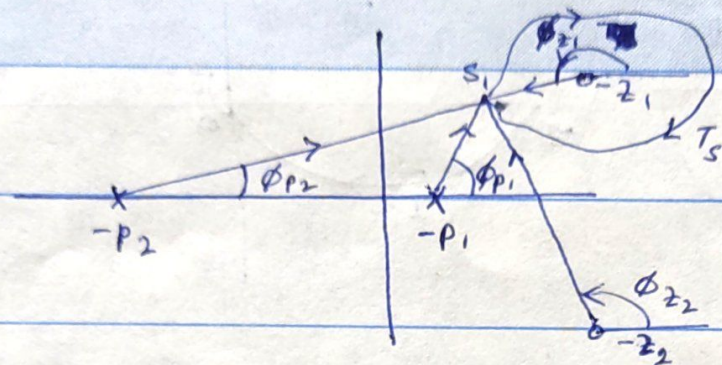
But if pt. of singularity or pole on $j\omega$ axis then small detour.

— to keep region enclosed in s -plane s.t. $F(s)$ is analytic — value exists



CAUCHY'S THEOREM :

$$\text{Say } F(s) = \frac{(s+z_1)(s+z_2)}{(s+p_1)(s+p_2)} = |F(s)| (\phi_{z_1} + \phi_{z_2} - \phi_{p_1} - \phi_{p_2})$$



\therefore As S_1 traverses T_5 , angles generated by $-p_1, -p_2, -z_2$ after one complete round are 0, while \angle generated by $-z_1$ is 2π radians (clockwise)

\therefore If Z zeros & P poles are encircled, then angle generated is $2\pi(Z) - 2\pi(P) = \phi_F = 2\pi N = \text{net angle generated by } T_F$ and hence net encirclement of origin of $F(s)$ plane in CW direction is $N = Z - P$.

Note $GH(s) + 1 = F(s) \Rightarrow GH(s)$ plane $(-1, 0)$ pt. $\equiv F(s)$ plane origin

\therefore Polar plot encirclement of $(-1, 0)$ pt. $\equiv F(s)$ plane encirclement of origin.

\therefore If there are P poles of GH in RHS and N encirclements of $(-1, 0)$
 $\Rightarrow Z$ of $F(s) = 1 + GH(s) = N + P$

Required to be zero for A. stable $\&$ system $\Rightarrow N = -P$ required.

NYQUIST STABILITY CRITERION :

- The Nyquist plot is obtained by mapping the fn. $GH(s)$ - polar plot.
- The Nyquist contour is chosen to enclose entire RHS of s -plane

Then $Z = N + P$

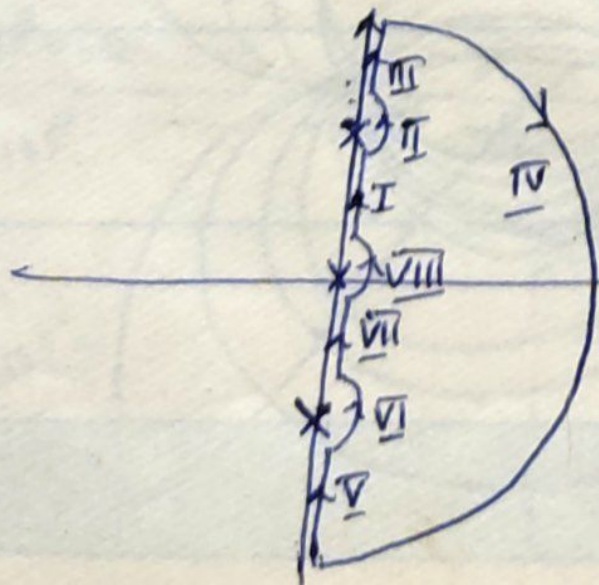
For $\&$ system to be stable, $Z = 0 \Rightarrow N = -P$.

Note: Since multiple loop systems may NOT include poles in RHS
 \therefore Routh's stability test must be performed.

Nyquist path for Q poles on $j\omega$ axis

$$GH(s) = \frac{K}{s(s^2 + \omega_1^2)(s+a)}$$

$$a > 0$$



I

$$s = j\omega$$

$$j0^+ \text{ to } j\omega_1^-$$

II

$$s = \epsilon t j\omega_1 + \epsilon e^{j\theta}$$

$$\theta \in [-\pi/2, \pi/2]$$

III

$$s = j\omega$$

$$j\omega_1^+ \text{ to } j\infty$$

IV

$$s = \epsilon t R e^{j\phi}$$

$$\phi \in [\pi/2, -\pi/2]$$

V to VII

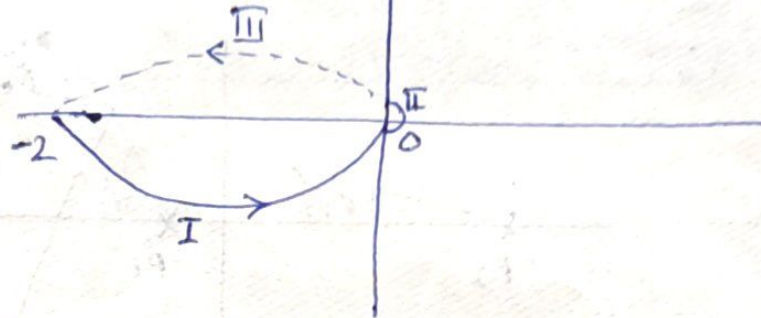
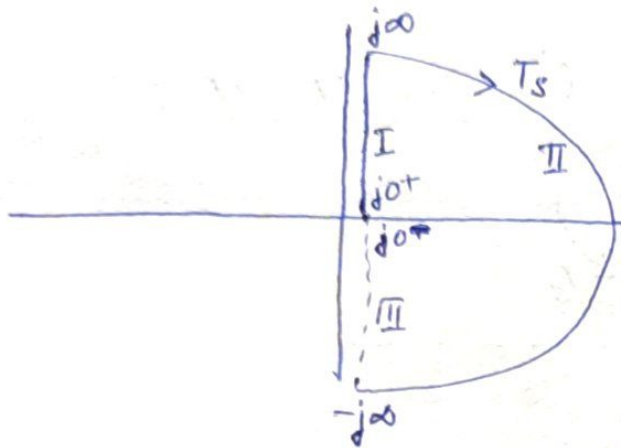
IPP of III to I

VIII

$$s = \epsilon e^{j\theta}$$

$$\theta \in [-\pi/2, \pi/2]$$

Ex 1: $G(s)H(s) = \frac{K(s+2)}{(s+1)(s-1)}$ where $K=1$.



I $G_H(s) = \frac{s+2}{(s+1)(s-1)}$

$\therefore |G_H(j\omega)| = \left| \frac{j\omega + 2}{(j\omega + 1)(j\omega - 1)} \right|$

Note: \angle varies from $+\pi/2$ to $-\pi/2$ through 0° on T_s semicircle of infinite radius

$\therefore \angle G_H = \angle \tan^{-1} \omega/2 - \angle \tan^{-1} \omega - \angle \pi - \tan^{-1} \omega = -\pi + \tan^{-1} \omega/2$
 for $\omega = 0$ to $\infty \rightarrow \angle G_H \in [-\pi, -\pi/2]$

II $s = \lim_{R \rightarrow \infty} R e^{j\phi} \therefore \lim_{R \rightarrow \infty} |G_H(s)| = \lim_{R \rightarrow \infty} \frac{1}{R} e^{j\phi} = 0 e^{j\phi}$

$\therefore \infty$ radius circle \odot CCW $\phi \in [\pi/2, -\pi/2]$ $\angle G_H \in [-\pi/2, \pi/2]$
CCW

III inverse plot of I.

$\therefore N = -1, P = +1 \therefore Z = N + P = 0 \therefore \& \text{ stable}$

✓ Ex 6 $G_H(s) = \frac{k(s+3)}{s(s-1)}$ $k=10$

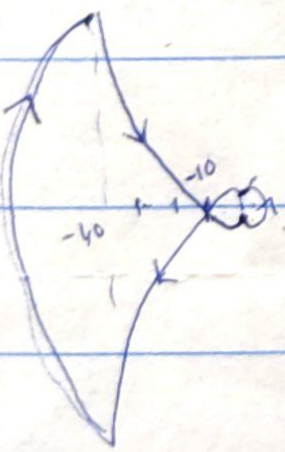


I $G_H(j\omega) = \frac{k(j\omega+3)}{j\omega(j\omega-1)} = \frac{k(j\omega+3)}{+j\omega(1+j\omega)} = \frac{-4k - jk(3\omega)}{1-\omega^2 - j\omega(1+\omega^2)}$

$\therefore \angle G_H(j\omega) = -\pi/2 + \tan^{-1} \frac{\omega}{3} - \tan^{-1}(\omega) \quad (\pi - \tan^{-1} \omega)$
 $= -3\pi/2 + \tan^{-1} \frac{\omega}{3} + \tan^{-1} \omega = \pi/2 + \tan^{-1} \frac{\omega}{3} + \tan^{-1} \omega$
 $(\tan^{-1} A + \tan^{-1} B) = \tan^{-1} \left(\frac{A+B}{1-AB} \right) \quad \left[\frac{3\pi}{2}, -\pi \right] \text{ CCW}$
 $= \left[\frac{\pi}{2}, \frac{3\pi}{2} \right] \text{ CW}$

\therefore For $\text{Im } GH = 0$, $\pi = \pi/2 + \tan^{-1} \left(\frac{\omega/3 + \omega}{1 - \omega^2/3} \right) \Rightarrow \omega = \sqrt{3}, |GH| = +10$

Ex 6



II $s = \rho e^{j\phi}$ $R \rightarrow \infty$ $\phi \in [\pi/2, -\pi/2]$ CW

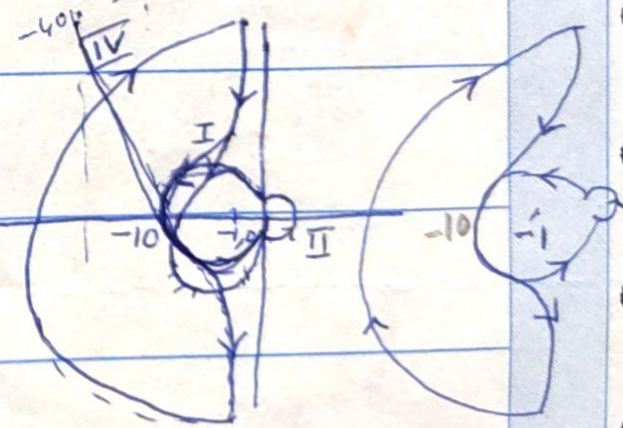
$G_H(s) = \frac{10}{R} e^{-j\phi}$ $\therefore \angle GH \in [-\pi/2, \pi/2]$ CCW

III IPP $\left[+\frac{\pi}{2}, +\frac{3\pi}{2} \right] \text{CCW}$

IV $s = \rho e^{j\theta}$ $\rho \rightarrow 0$ $\theta \in [-\pi/2, \pi/2]$ CCW

$G_H(s) = \frac{30}{-s} \Big|_{s = \rho e^{j\theta}} = (-1) \cdot \frac{30}{\rho} = \infty \angle +\pi - \theta$

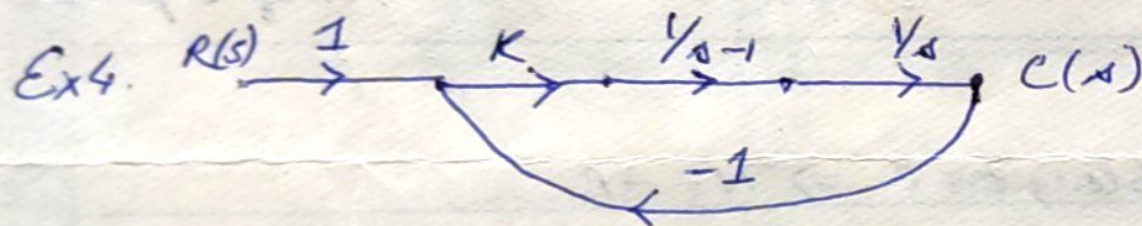
$\therefore \angle GH = \left[+\frac{3\pi}{2}, +\frac{\pi}{2} \right] \text{CW}$



$\therefore P = 1, N = -1 \therefore Z = 0$ STABLE.

Ex 2 $GH(s) = \frac{K}{s(s+a)}$

Ex 3 $GH(s) = \frac{K(T_2s+1)}{s^2(T_1s+1)}$ $T_1 > T_2$



Ex 5. $GH(s) = \frac{2(s+0.1)(s+0.6)(s^2+s+1)}{s^3(s-0.2)(s+1)}$

Ex 6. $GH(s) = \frac{K(s+3)}{s(s-1)}$ $K=10$

Ex 7. $GH(s) = \frac{K(s-1)}{s(s+1)}$ $K > 0$. Check roots of $F(s)$

Ex 8. $GH(s) = \frac{K}{s(T_1s+1)(T_2s+1)}$

Conditionally stable systems a) $GH(s) = \frac{K(s^2+2s+4)}{s(s+4)(s+6)(s^2+1.1s+1)}$

b) $GH(s) = \frac{100K(s+5)(s+40)}{s^3(s+100)(s+200)}$