# **Process Controllers**

Prof. Anjan Rakshit and Dr. Amitava Chatterjee Electrical Measurements and Instrumentation Laboratory, Electrical Engineering Department, Jadavpur University, Kolkata, India.









Realization of Electronic PID controller (P, PI, & PD are special cases of PID)

The controller output may be expressed as:

$$m = K_p \left( e + \frac{1}{T_i} \int_0^t e dt + T_D \frac{de}{dt} \right)$$

where  $T_i$  = integral time constant  $T_D$  = derivative time constant

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or, 
$$m = K_p e + \frac{K_p}{T_i} \int_0^t e dt + K_p T_D \frac{de}{dt}$$

Taking Laplace transform with zero initial conditions,

$$M(s) = K_{p}E(s) + \frac{K_{p}E(s)}{sT_{i}} + sK_{p}T_{D}E(s)$$

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or, 
$$\frac{M(s)}{E(s)} = K_{p}\left(1 + \frac{1}{sT_{i}} + sT_{D}\right)$$

the transfer function of PID controller

#### Parallel Realization of PID Controller







 $M(s) = -(V_{1}(s) + V_{2}(s) + V_{3}(s))$ 



$$M(s) = -(V_{1}(s) + V_{2}(s) + V_{3}(s))$$
  
or, 
$$M(s) = E(s) \left[ \frac{R_{1}}{R_{2}} + \frac{1}{sR_{I}C_{I}} + sR_{D}C_{D} \right]$$

Therefore, transfer function of PID controller becomes

$$\frac{M(s)}{E(s)} = \left[\frac{R_1}{R_2} + \frac{1}{sR_1C_1} + sR_DC_D\right]$$

Here,

$$K_{p} = \frac{R_{1}}{R_{2}}$$
$$\frac{K_{p}}{T_{i}} = \frac{1}{R_{I}C_{I}}$$
$$K_{p}T_{D} = R_{D}C_{D}$$

**Proportional and integral terms may be combined as:** 



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$$\frac{E(s)}{R_{2}} + \frac{V_{1}(s)}{R_{1}} = 0$$
  
or, 
$$\frac{E(s)}{R_{2}} = -\frac{V_{1}(s)}{R_{1} + \frac{1}{sC_{I}}} = -\frac{V_{1}(s)sC_{I}}{1 + sR_{1}C_{I}}$$

Or, 
$$\frac{V_1(s)}{E(s)} = -\frac{1 + sR_1C_1}{sR_2C_1}$$

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$$\frac{V_{1}(s)}{E(s)} = -\frac{1 + sR_{1}C_{1}}{sR_{2}C_{1}}$$
  
=  $-\left(\frac{R_{1}}{R_{2}} + \frac{1}{sR_{2}C_{1}}\right)$ 

Or, 
$$\frac{V_1(s)}{E(s)} = -\frac{1 + sR_1C_1}{sR_2C_1}$$
  
=  $-\left(\frac{R_1}{R_2} + \frac{1}{sR_2C_1}\right)$   
=  $-\frac{R_1}{R_2}\left(1 + \frac{1}{sR_1C_1}\right)$ 

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$$\frac{V_{1}(s)}{E(s)} = -\frac{1 + sR_{1}C_{1}}{sR_{2}C_{1}}$$
  
 $= -\left(\frac{R_{1}}{R_{2}} + \frac{1}{sR_{2}C_{1}}\right)$   
 $= -\frac{R_{1}}{R_{2}}\left(1 + \frac{1}{sR_{1}C_{1}}\right)$   
 $= -K_{p}\left(1 + \frac{1}{sT_{i}}\right)$ 

Where

$$K_{p} = \frac{R_{1}}{R_{2}}$$
 and  $T_{i} = R_{1}C_{I}$ 

A simple PID Controller with two op-amps



A simple PID Controller with two op-amps



A simple PID Controller with two op-amps



The unity gain buffer amplifier is required to avoid the loading effect of the feedback network





Loop equations:

$$\beta M(s) = I_{1} \left( R_{I} + \frac{1}{sC_{I}} \right) - I_{2}R_{I}$$
$$0 = -I_{1}R_{I} + I_{2} \left( R_{I} + R_{D} + \frac{1}{sC_{D}} \right)$$



Loop equations:

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In matrix form:

$$\begin{bmatrix} R_I + \frac{1}{sC_I} & -R_I \\ -R_I & R_I + R_D + \frac{1}{sC_D} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} \beta M(s) \\ 0 \end{bmatrix}$$



By Cramer's rule:

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$$I_2 = \frac{\begin{vmatrix} R_I + \frac{1}{sC_I} & \beta M(s) \\ -R_I & 0 \end{vmatrix}}{\begin{vmatrix} R_I + \frac{1}{sC_I} & -R_I \\ -R_I & R_I + R_D + \frac{1}{sC_D} \end{vmatrix}}$$





Or, 
$$I_2 = \frac{\beta M(s)R_I}{\left(R_I + \frac{1}{sC_I}\right)\left(R_I + R_D + \frac{1}{sC_D}\right) - R_I^2}$$



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Or, 
$$I_2 = \frac{\beta M(s) R_I}{\left(R_I + \frac{1}{sC_I}\right)\left(R_I + R_D + \frac{1}{sC_D}\right) - R_I^2}$$
  
$$\therefore E(s) = \frac{I_2}{sC_D} = \frac{\beta M(s) R_I}{\left(R_I + \frac{1}{sC_I}\right)\left(R_I + R_D + \frac{1}{sC_D}\right) - R_I^2}$$



Then, the T.F. becomes:

$$\frac{M(s)}{E(s)} = \frac{1}{\beta} \left[ \frac{\left\{ \left(R_I + \frac{1}{sC_I}\right) \left(R_I + R_D + \frac{1}{sC_D}\right) - R_I^2 \right\} sC_D}{R_I} \right]}{R_I} \right]$$

Then, the T.F. becomes:



Simplifying,

e

C<sub>D</sub>:

R<sub>D</sub>

$$\frac{M(s)}{E(s)} = \frac{1}{\beta} \left[ \left( \frac{T_1 + T_2 + R_I C_D}{T_1} \right) + \frac{1}{sT_1} + sT_2 \right]$$

where  $T_1 = R_I C_I$  and  $T_2 = R_D C_D$ 

▶ m

βm



By substituting

$$A = \frac{T_1 + T_2 + R_I C_D}{T_1}$$

Or, 
$$A = 1 + \frac{C_D}{C_I} + \frac{T_2}{T_1}$$



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This expression of A presents the problem of interaction and hence, the controller developed is called an **interacting controller**.



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**Or**, 
$$A = 1 + \frac{C_D}{C_I} + \frac{T_2}{T_1}$$

The T.F. becomes:

$$\frac{M(s)}{E(s)} = \frac{A}{\beta} \left[ 1 + \frac{1}{sT_1A} + \frac{sT_2}{A} \right]$$
$$= K_p \left[ 1 + \frac{1}{sT_I} + sT_D \right]$$
Where  $K_p = \frac{A}{\beta}, T_I = T_1A$ , and  $T_D = T_2/A$ 

A simplified PID Controller with one op-amp



The unity gain buffer amplifier may be omitted if the resistances are chosen such that  $(R_{_I} \parallel R_{_D}) >> (R_{_1} \parallel R_{_2})$
A simplified PID Controller with one op-amp



The unity gain buffer amplifier may be omitted if the resistances are chosen such that

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Provision for providing a bias term

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Provision for anti derivative kick

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Provision for anti-integral wind-up

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- Provision for Local/Remote modes of operation

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#### Provision for anti-derivative kick

Derivative action may produce an unwanted kick at the controller output when there is a step change in the set-point. This effect may be eliminated if the derivative term is computed from the measured variable, instead of the error (anti-derivative kick feature).

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PD-control with derivative kick

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PD-control with derivative kick



PD-control with anti-derivative kick

#### Provision for anti-integral wind-up

Integral action in PI and PID controllers may produce a large integral error term when a non-zero error persists for a long time. Integration action may be switched off to combat this situation, otherwise, a long time may be required to come back to normal working range.

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In some controllers, when the integral term is present, the integral amount during  $T_i$  is summed with each pass through the calculation and becomes the controller bias. This technique is known as automatic reset and this is done to avoid long duration operation of the integral action – thus error due to drift etc. during integration may be avoided.

# Output must be limited (saturation)

It is desirable to limit the controller output (say between 0% and 100%) so that control valves or other final control elements may operate safely within their working limits.

# Provision for Auto/Manual modes of control

A change-over switch is normally provided for configuring the controller as an automatic controller (AUTO) for closed loop operation or as a manual controller (MAN) for open-loop operation.

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# Provision for Local/Remote modes of operation

A provision is made for the set point input, such that, the set point may be changed either locally or from a remote link.

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> No transmission lag (as compared to pneumatic systems)

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- > Compatibility with other electrical components

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> Analog integrators are not very reliable

> Conversion equipments are necessary to interface pneumatic and hydraulic devices









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x: change in baffle-nozzle separation. m: change in back pressure.  $P_a$ : the lowest possible pressure ( = ambient pressure).  $X = X^2$ , x and  $M = M^2$ , m, X<sup>2</sup>; baffle pozzle concretion with zero.

 $X = X^{-} x$  and  $M = M^{+} m$ , X<sup>+</sup>: baffle-nozzle separation with zero error and M<sup>+</sup>: output pressure with zero error.





For approximately linear working range:  $M = K_n X + C$  and  $M^{\sim} = K_n X^{\sim} + C$ ,  $K_n$  being the slope



For approximately linear working range:  $M = K_n X + C$  and  $M^{`} = K_n X^{`} + C$ ,  $K_n$  being the slope Subtracting,  $M - M^{`} = K_n(X - X^{`})$ , i.e.,  $m = -K_n \cdot X$ 



Under steady state condition, change in output pressure may be expressed as:

$$m = -K_n \cdot x$$
, where  $-K_n$  is the nozzle gain  
Here,  $-x = \left(\frac{a}{a+b}\right)e$
#### Pneumatic Baffle-Nozzle or Flapper-Nozzle amplifier



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Here,  $-x = \left(\frac{a}{a+b}\right)e$   
Thus,  $m = K_n \cdot e\left(\frac{a}{a+b}\right) = K \cdot e$ , *K* is called the amplifier gain and  $K = K_n\left(\frac{a}{a+b}\right)$ 

# Arrangement for mechanical set-point and pneumatic measured variable



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**Bellows** 



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A relay valve may be *direct-acting* (positive gain) or *reverse-acting* (negative gain).

#### Direct-acting relay valve with a Baffle-Nozzle amplifier



As the nozzle back pressure *m* increases, the relay output pressure also increases

# Direct-acting relays



#### **Bleed type relay**

In all positions of the valve, except at the position to shut off the air supply, air continues to bleed into the atmosphere.

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**Bleed type relay** 



#### Non-bleed type relay

The air bleed stops when the equilibrium condition is obtained, and, therefore, there is no loss of pressurized air at steady-state operation.

# Reverse-acting relay



As the nozzle back pressure increases, the ball valve is forced towards the lower seat, thereby decreasing the output pressure





The Baffle-Nozzle separation may be expressed as:

$$-(x) = \left(\frac{a}{a+b}\right)e - \left(\frac{b}{a+b}\right)K_bm$$

where  $K_b$  is the bellows stiffness factor and m is the change in output pressure.



$$-(x) = \left(\frac{a}{a+b}\right)e - \left(\frac{b}{a+b}\right)K_bm$$

Now  $m = -K_n . x$ , where  $K_n$  is the nozzle gain.



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#### **Block diagram of the controller**





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Thus, 
$$m/K_n = \left(\frac{a}{a+b}\right)e - \left(\frac{b}{a+b}\right)K_bm$$

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or  $m\left[\frac{1}{K_n} + \left(\frac{b}{a+b}\right)K_b\right] = \left(\frac{a}{a+b}\right)e$ 



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or  $m\left[\frac{1}{K_n} + \left(\frac{b}{a+b}\right)K_b\right] = \left(\frac{a}{a+b}\right)e$   
if  $K_n$  is very high, then  $\frac{1}{K_n} \approx 0$ ,



This gives

$$m = \frac{\left(\frac{a}{a+b}\right)}{\left(\frac{b}{a+b}\right)K_b}e = \left(\frac{a}{bK_b}\right)e = K_p.e$$

where  $K_p$  = proportional gain =

$$\frac{a}{bK_{b}}$$



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#### Simplified block diagram of the controller

$$E(s) \rightarrow K_p \rightarrow M(s)$$

Pneumatic proportional controller with a direct-acting relay







The Baffle-Nozzle separation may be expressed as:

$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)\left(\frac{K_b}{1+sT_D}\right)M(s) \quad \text{(assuming } R_D >> R)$$

where  $K_b$  = Bellows stiffness factor,  $T_D$  = Derivative time  $= R_D C_D$ , (assuming  $R_D >> R$ ),  $C_D$  = Capacity of the bellows.



$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)\left(\frac{K_b}{1+sT_D}\right)M(s)$$

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where  $K_n$  is the nozzle gain.

Thus 
$$\frac{M(s)}{K_n} = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{b}{a+b}\right) \left(\frac{K_b}{1+sT_D}\right) M(s)$$



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or, 
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 Now,  $K_n >> 1, \frac{1}{K_n} \approx 0,$ 



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or, 
$$M(s) \left[\frac{1}{K_n} + \left(\frac{b}{a+b}\right) \left(\frac{K_b}{1+sT_D}\right)\right] = \left(\frac{a}{a+b}\right) E(s) \quad \text{Now,} \quad K_n \gg 1, \frac{1}{K_n} \approx 0,$$

$$\therefore \frac{M(s)}{E(s)} \approx \left(\frac{a}{bK_b}\right) (1+sT_D)$$

$$= K_p (1+sT_D) \quad \text{where } K_p = \text{proportional gain} = \frac{a}{bK_b}$$





**The Baffle-Nozzle separation may be expressed as:** (assuming R<sub>1</sub> >> R and same stiffness for both bellows)

$$X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)K_{b}M(s) + \left(\frac{b}{a+b}\right)\left(\frac{K_{b}}{1+sT_{i}}\right)M(s)$$

where  $T_i = Integral time$ =  $R_IC_I$ and  $C_I = Capacity of integral bellows$  $<math>K_b = Bellows stiffness factor.$ 



$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)K_{b}M(s) + \left(\frac{b}{a+b}\right)\left(\frac{K_{b}}{1+sT_{i}}\right)M(s)$$

Hence,

$$-X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(1 - \frac{1}{1+sT_i}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{1}{1+\frac{1}{sT_i}}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$



$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)K_bM(s) + \left(\frac{b}{a+b}\right)\left(\frac{K_b}{1+sT_i}\right)M(s)$$

Hence,

$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(1 - \frac{1}{1+sT_i}\right)\left(\frac{b}{a+b}\right)K_bM(s)$$
$$= \left(\frac{a}{a+b}\right)E(s) - \left(\frac{1}{1+\frac{1}{sT_i}}\right)\left(\frac{b}{a+b}\right)K_bM(s)$$

Now,  $M(s) = -K_n \cdot X(s)$ 



$$X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(1 - \frac{1}{1+sT_i}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{1}{1+\frac{1}{sT_i}}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$
and 
$$M(s) = -K_n X(s)$$

Thus, assuming  $K_n >> 1$ , i.e.  $\frac{1}{K_n} \approx 0$ ,



$$-X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(1 - \frac{1}{1+sT_i}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{1}{1+\frac{1}{sT_i}}\right) \left(\frac{b}{a+b}\right) K_b M(s)$$
and  $M(s) = -K_n \cdot X(s)$ 

Thus, assuming  $K_n >> 1$ , i.e.  $\frac{1}{K_n} \approx 0$ ,

$$\frac{M(s)}{E(s)} = \frac{a}{bK_b} \left(1 + \frac{1}{sT_i}\right) = K_p \left(1 + \frac{1}{sT_i}\right), \text{ where } K_p = \frac{a}{bK_b}$$




The Baffle-Nozzle separation may be expressed as:

(assuming  $R_1 >> R_D >> R$ )

$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{b}{a+b}\right)\left(\frac{K_{b}}{1+sT_{D}}\right)M(s)$$
$$+ \left(\frac{b}{a+b}\right)\left(\frac{K_{b}}{(1+sT_{i})(1+sT_{D})}\right)M(s)$$
$$= \left(\frac{a}{a+b}\right)E(s) - \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{(1+sT_{D})} - \frac{1}{(1+sT_{i})(1+sT_{D})}\right)M(s)$$



$$X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{1+sT_{D}}\right) M(s)$$
$$+ \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_{b}}{a+b}\right) \left(\frac{1}{(1+sT_{D})} - \frac{1}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$

Now,  $\frac{1}{1+sT_{D}} - \frac{1}{(1+sT_{i})(1+sT_{D})} = \frac{1+sT_{i}-1}{(1+sT_{i})(1+sT_{D})}$ 



$$X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{1+sT_{D}}\right) M(s)$$
$$+ \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_{b}}{a+b}\right) \left(\frac{1}{(1+sT_{D})} - \frac{1}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$

Now,  

$$\frac{1}{1+sT_{D}} - \frac{1}{(1+sT_{i})(1+sT_{D})} = \frac{1+sT_{i}-1}{(1+sT_{i})(1+sT_{D})}$$

$$= \frac{sT_{i}}{1+sT_{i}+sT_{D}+s^{2}T_{i}T_{D}}$$

$$= \frac{1}{1+\frac{1}{sT_{i}}+\frac{T_{D}}{T_{i}}+sT_{D}}$$



$$X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{1+sT_{b}}\right) M(s)$$
$$+ \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{(1+sT_{i})(1+sT_{b})}\right) M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_{b}}{a+b}\right) \left(\frac{1}{(1+sT_{b})} - \frac{1}{(1+sT_{i})(1+sT_{b})}\right) M(s)$$

Therefore,

$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1+\frac{1}{sT_{i}}+\frac{T_{b}}{T_{i}}+sT_{b}}\right)M(s)$$



$$X(s) = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{1+sT_{D}}\right) M(s)$$
$$+ \left(\frac{b}{a+b}\right) \left(\frac{K_{b}}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$
$$= \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_{b}}{a+b}\right) \left(\frac{1}{(1+sT_{D})} - \frac{1}{(1+sT_{i})(1+sT_{D})}\right) M(s)$$

Therefore,

$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1+\frac{1}{sT_{i}}+\frac{T_{b}}{T_{i}}+sT_{b}}\right)M(s)$$

Now,

$$M(s) = -K_n \cdot X(s)$$



$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1+\frac{1}{sT_{i}}+\frac{T_{b}}{T_{i}}+sT_{b}}\right)M(s)$$

and  $M(s) = -K_n \cdot X(s)$ 

Therefore,

$$\frac{M(s)}{K_n} = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_b}{a+b}\right) \left(\frac{1}{1+\frac{T_b}{T_i}+\frac{1}{sT_i}+sT_b}\right) M(s)$$



$$-X(s) = \left(\frac{a}{a+b}\right)E(s) - \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1+\frac{1}{sT_{i}}+\frac{T_{b}}{T_{i}}+sT_{b}}\right)M(s)$$

and  $M(s) = -K_n \cdot X(s)$ 

Therefore,

$$\frac{M(s)}{K_n} = \left(\frac{a}{a+b}\right) E(s) - \left(\frac{bK_b}{a+b}\right) \left(\frac{1}{1+\frac{T_b}{T_i} + \frac{1}{sT_i} + sT_b}\right) M(s)$$
  
or, 
$$M(s) \left[\frac{1}{K_n} + \left(\frac{bK_b}{a+b}\right) \left(\frac{1}{1+\frac{T_b}{T_i} + \frac{1}{sT_i} + sT_b}\right)\right] = \left(\frac{a}{a+b}\right) E(s)$$



$$M(s)\left[\frac{1}{K_{n}} + \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1 + \frac{T_{D}}{T_{i}} + \frac{1}{sT_{i}} + sT_{D}}\right)\right] = \left(\frac{a}{a+b}\right)E(s)$$



$$M(s)\left[\frac{1}{K_{n}} + \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1 + \frac{T_{D}}{T_{i}} + \frac{1}{sT_{i}} + sT_{D}}\right)\right] = \left(\frac{a}{a+b}\right)E(s)$$
  
as  $K_{n} \gg 1$ ,  $\frac{1}{K_{n}} \approx 0$ 



$$M(s)\left[\frac{1}{K_{n}} + \left(\frac{bK_{b}}{a+b}\right)\left(\frac{1}{1 + \frac{T_{D}}{T_{i}} + \frac{1}{sT_{i}} + sT_{D}}\right)\right] = \left(\frac{a}{a+b}\right)E(s)$$
  
as  $K_{n} >> 1$ ,  $\frac{1}{K_{n}} \approx 0$ 

Therefore,

$$\frac{M(s)}{E(s)} = \left(\frac{a}{bK_{b}}\right) \left(1 + \frac{T_{D}}{T_{i}} + \frac{1}{sT_{i}} + sT_{D}\right) = K_{p} \left(1 + \frac{T_{D}}{T_{i}} + \frac{1}{sT_{i}} + sT_{D}\right)$$

where

$$K_{p} = \frac{a}{bK_{b}}$$







$$\frac{M(s)}{E(s)} = \left(\frac{a}{bK_b}\right) \left(1 + \frac{T_D}{T_i} + \frac{1}{sT_i} + sT_D\right) = K_p \left(1 + \frac{T_D}{T_i} + \frac{1}{sT_i} + sT_D\right) \quad \text{where} \quad K_p = \frac{a}{bK_b}$$

Here,



imposes an interaction between integral and derivative operations of the controller. If we choose  $T_i >> T_D$ , the interaction reduces and the transfer function becomes

$$\frac{M(s)}{E(s)} \approx K_p \left( 1 + \frac{1}{sT_i} + sT_D \right) \quad \text{the ideal relation of a PID controller.}$$



The controller gain becomes infinite when  $T_i = T_D$ , if derivative time restriction is placed in position (1).

#### References

- 1. Process Control Systems by Shinskey
- 2. Automatic Process Control by Eckman
- 3. Principles of Process Control by Patranabis
- 4. Process Control by Harriott
- 5. Process Systems Analysis and Control by Coughanowr and Koppel
- 6. Process Control by Pollard
- 7. Chemical Process Control by Stephanopoulos
- 8. Modern Control Engineering by Ogata
- 9. Applied Process Control by Chidambaram

